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INPUT FREQUENCY REQUIREMENTS FOR IDENTIFICATION THROUGH LIAPUNOV METHODS

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I. Introduction

In this paper a theorem is derived which specifies a sufficient number of input frequencies to guarantee identification of an unknown noise free linear plant. Similar work has been reported by Lion [1], Carroll [2], and Hoberock and Stewart [3], in which various classes of systems were considered. The results in this paper, although essentially in agreement with previous work, are in some respects more general. Since all of the referenced work relates to sufficient conditions for nulling of the parameter error vector, it is to be expected that various conditions on the system to be identified will have been imposed. In this paper it is found that some of these conditions appear to be necessary while others do not. The main contribution is to provide a theorem which considers the effect of unknown parameters in the state equation upon the frequency requirements.

Results are obtained for a linear multi-input system of known structure in which some or all of the coefficients of the state equation are unknown. As in [1],[2], an identification algorithm is assumed by which convergence of the parameter error vector is guaranteed, and requirements on the input frequencies are derived such that convergence to the null is assured. In [1] it was specified that the plant transfer function has a single input, and contains no phase shifts of integral multiples of π radians at the input frequencies. In [2] observability and controllability, but no phase shift requirements were imposed. In [4], conditions for identifiability of the multi-input plant were studied without regard to a specific identification algorithm. The conditions imposed on the plant were rather stringent, however, in that no eigenvalues on the imaginary axis were permitted.

In the present work the identification scheme outlined by Kudva and Narendra [4] is used as a model. However, the theorem is applicable to the model following adaptive control problems and adaptive observers as well.

Simulation results are presented which demonstrate certain interesting aspects of the theorem. It is seen that the sufficient number of input frequencies can be quite small relative to the system order, depending upon the location and number of the unknown coefficients of the state equation. It is further demonstrated that convergence of the parameter error vector to the null may not occur when the number of frequencies fails to satisfy the theorem. Although certain phase shift requirements appear in the theorem, as in [1], it is shown that convergence to the null may in fact occur when these requirements are violated. Finally, it is demonstrated that the assumption of controllability made in the present and referenced theorems is not a necessary condition.

II. Problem Formulation

The mathematical expressions of plant and model are described as follows:

PLANT:
$$\dot{x} = A \dot{x} + B \dot{u}$$
, (A*, B* constant, unknown) (1)

MODEL:
$$\dot{x}_m = Ax + C[x_m - x] + Bu$$
 (2)

where A, A*eR^{nxn}; B, B*eR^{nxm}; x_m, xeRⁿ; ueR^m; and C is any stability matrix. It is assumed that all the plant states are measurable and the dimensions of A* and B* are also known. The purpose here is to identify the plant by adjusting the parameters of A and B.

Now defining the tracking error as

$$e \stackrel{\Delta}{=} x_{m} - x \tag{3}$$

the following error equation is obtained:

$$\dot{\mathbf{e}} = \mathbf{C}\mathbf{e} + \mathbf{\Phi}\mathbf{x} + \mathbf{\Psi}\mathbf{u} \tag{4}$$

where the parameter errors are defined by the matrices

$$\Phi = A - A^*$$

$$\Psi = B - B^*$$

Following the procedure shown in [4], it is possible to obtain a Liapunov function which is a positive definite quadratic form in tracking and parameter error space,

$$V = \frac{1}{2} (e^{T} P e + \sum_{i=1}^{n} \phi_{i}^{T} \phi_{i} + \sum_{i=1}^{m} \psi_{i}^{T} \psi_{i})$$
 (5)

where

$$P = P^{T}$$
 is positive definite
 $\phi_{i} \stackrel{\Delta}{=} i^{th}$ column of Φ
 $\psi_{i} \stackrel{\Delta}{=} i^{th}$ column of Ψ

The time derivative is in turn given by

$$\dot{\mathbf{V}} = -\mathbf{e}^{\mathrm{T}}\mathbf{Q}\mathbf{e} + \mathbf{e}^{\mathrm{T}}\mathbf{P}\mathbf{f} + \sum_{i=1}^{n} \dot{\phi}_{i}^{\mathrm{T}}\phi_{i} + \dot{\psi}_{i}^{\mathrm{T}}\psi_{i}$$
 (6)

where

$$-Q \stackrel{\Delta}{=} (C^{T}P + PC)/2 , Q = Q^{T}$$

$$f \stackrel{\Delta}{=} \Phi x + \psi u$$

and Q is any positive definite symetric matrix. If the adaptive law is formulated as

In this report the problem is formulated as one of identification. However, the results can also be applied to the adaptive control problem as outlined in [5]. The term "adaptive system" will be used to imply either "model reference adaptive control" or "parameter identification".

then (6) becomes

$$\dot{\mathbf{v}} = -\mathbf{e}^{\mathrm{T}}\mathbf{0}\mathbf{e} \tag{9}$$

which is seen to be semi-negative definite in the e, Φ , Ψ space. (5) and (9) show that there is an uncertainty as to the asymptotic behavior of the Φ and Ψ matrices, in the sense that, although Φ , Ψ are assured of attaining constant values, there is no guarantee that they will approach the null.

It is shown in the next section that if a certain class of inputs is applied, then the convergence of Φ , Ψ to the null can be guaranteed.

III. Main Result (Single Input Case)

As a solution to this problem, the following theorem is established. Theorem:

In a single input adaptive system, the convergence of the Φ , Ψ matrices of (4) to the null is guaranteed if

- 1) the input is periodic and contains at least q distinct frequencies where q = N.U.I.[R/2] and $R \le n + 1,^{2,3}$
- 2) there exists at least one x in which none of the q distinct frequencies encounters a phase shift of $K\pi/2$ radians, where K is any integer,
- a) for all non-zero columns of Φ in (10) the corresponding states are excited and linearly independent in the steady state 4 at the input frequencies.

Proof:

Since Liapunov adaptation quarantees asymptotic stability of the equilibrium in e, then for $t\to\infty$, $e\equiv\dot{e}\equiv0$. Thus (4) in the steady state becomes

$$\Phi x + \Psi u = 0 \tag{10}$$

 $[\]frac{\Delta}{2}$ N.U.I. = NEAREST UPPER INTEGER.

³Let R_i = the sum of number of nonzero elements in ith row of Φ , Ψ . Then R = U.B.[R_i], i = 1,2,...n.

⁴By this requirement, the system must also be controllable.

Assuming the ith row of the above equation contains R nonzero elements, and letting R be equal to (n+1), the ith equation in (10) is given by⁵

$$\sum_{j=1}^{n} \phi_{ij} x_{j} + \psi_{i} u = 0 , \quad i = 1, 2, ... n$$
 (11)

where the input, u, is assumed to be comprised of q sinusoidal signals with distinct frequencies, namely

$$u = \sum_{\ell=1}^{q} u^{\ell}$$
 (12)

where

$$q = N.U.I.[R/2]$$
 $u^{\ell} = \sin \omega_0 t$

Now applying the superposition property to (11), a set of q equations is obtained,

$$\sum_{j=1}^{n} \phi_{ij} x_{j}^{\ell} + \psi_{i} u^{\ell} = 0 , \quad \ell = 1, 2, \dots, q$$
 (13)

where x^{ℓ} is the state response to u^{ℓ} . Since u^{ℓ} is a periodic function, and the state response, x^{ℓ} , is also periodic in the steady state, all the terms in (13) are periodic. Because of the nature of the periodic input, the time derivative of (13) yields another independent equation, namely

$$\sum_{j=1}^{n} \phi_{ij} \dot{x}_{j}^{\ell} + \psi_{i} \dot{u}^{\ell} = 0 , \quad \ell = 1, 2, \dots, q$$
 (14)

wherein, by Liapunov stability, $\dot{\phi}_{ij} \equiv \dot{\psi}_{i} \equiv 0$ in the steady state for all j. The 2q (i.e. n+1) equations represented by (13) and (14) can be written in matrix form,

$$[F(t)]\gamma_i = 0 \tag{15}$$

⁵The case of R < (n+1), which is a subset of the case of R = n + 1, is not treated explicitly in the proof because of the tedious indexing required. An example is provided to show the case of R < (n+1).

where

$$F(t) = \begin{bmatrix} x_1^1 & x_2^1 & \dots & x_n^1 & u^1 \\ x_1^1 & x_2^1 & \dots & x_n^1 & u^1 \\ \vdots & \vdots & \vdots & \vdots \\ x_1^k & x_2^k & x_j^k & x_n^k & u^k \\ \vdots & \vdots & \vdots & \vdots \\ x_1^k & x_2^k & x_j^k & x_n^k & u^k \\ \vdots & \vdots & \vdots & \vdots \\ \frac{n+1}{2} & \frac{n+1}{2} & \frac{n+1}{2} & \frac{n+1}{2} \\ x_1 & x_2^2 & \dots & x_n^k & u^k \\ \vdots & \vdots & \vdots & \vdots \\ \frac{n+1}{2} & \frac{n+1}{2} & \frac{n+1}{2} & \frac{n+1}{2} \\ \frac{n+1}{2} & \frac{n+1}{2} & \frac{n+1}{2} & \frac{n+1}{2} \\ \frac{n+1}{2} & \frac{n+1}{2} & \frac{n+1}{2} & \frac{n+1}{2} \end{bmatrix}$$

in which

$$x_j^k = \left| K_j^k (j\omega_k) \right| \sin (\omega_k t + \alpha_j^k)$$

and the vector γ_i is defined by

$$\gamma_{i} \stackrel{\Delta}{=} [\phi_{i1}\phi_{i2} \dots \phi_{in}\psi_{i}]^{T}.$$

Using theorem (5-3) in [6] it is to be proved that $\gamma_i \equiv 0$ for all t or, what is equivalent, that F(t) is a non-singular matrix for all t.

Comment: Because F(t) is time varying it can be shown by example that linear independence of its rows (columns) need not infer linear independents of its columns (rows).

From this comment it is seen that linear independence of both rows and columns must be established in order to establish that F(t) is non-singular for all t. Since, as can be readily demonstrated, F(t) in (15) is identical for all $i=1,2,\ldots n$, once F(t) is proved to be nonsingular it follows that $\gamma_i\equiv 0$ for all $i=1,2,\ldots m$.

The linear independence of the rows of F(t) will be discussed first. If F(t) consists of analytic functions on the interval $[t_1, t_2]$ then by Theorem (5-3) if the rank of the infinite matrix

$$|\overline{F}(t_0)| : F^{(1)}(t_0) : F(t_0)^{(2)} : \dots : F^{(n-1)}(t_0) : \dots$$
(16)

is n+1, where $F^{(i)}(t)$ is the i^{th} derivative of F(t), F(t) is an (n+1)x(n+1) matrix, and t_0 is any fixed point in $[t_1,t_2]$, then the rows of F(t) are linearly independent for all $t \in [t_1,t_2]$.

Since F(t) in (15) consists of analytic functions, without loss of generality let $[t_1, t_2]$ be taken as $[0, \infty]$. Now setting $t_0=0$, and selecting the appropriate (n+1) columns from (16), an (n+1)x(n+1) matrix can be formed which has the following factored form:

$$\begin{bmatrix} 1 & -\omega_{1}^{2} & \omega_{1}^{4} & \dots & -\omega_{1}^{2(n+1)} & (K_{h}^{1} \sin \alpha_{h}^{1}/K_{j}^{1} \sin \alpha_{j}^{1}) \\ \omega_{1} & -\omega_{1}^{3} & \omega_{1}^{5} & \dots & \omega_{1}^{2n-1} & \omega_{1}(K_{h}^{1} \cos \alpha_{h}^{1}/K_{j}^{1} \cos \alpha_{j}^{1}) \\ 1 & -\omega_{2}^{2} & \omega_{2}^{4} & \dots & -\omega_{2}^{2(n-1)} & (K_{h}^{2} \sin \alpha_{h}^{2}/K_{j}^{2} \sin \alpha_{j}^{2}) \\ \omega_{2} & -\omega_{2}^{3} & \omega_{2}^{5} & \dots & \omega_{2}^{2n-1} & \omega_{2}(K_{h}^{2} \cos \alpha_{h}^{2}/K_{j}^{2} \cos \alpha_{j}^{2}) \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ \omega_{n+1} & -\omega_{n+1}^{3} & \omega_{n+1}^{5} & \dots & \omega_{n+1}^{2n-1} & \omega_{n+1}^{2} & (K_{h}^{2} \cos \alpha_{h}^{2}/K_{j}^{2} \cos \alpha_{h}^{2}) \end{bmatrix}$$

To begin with, pick jth column of (16), where $j \le n$, and then pick every $[j + k(n+1)]^{th}$ column where $k=1,2,\ldots n-1$. Finally add any column (e.g. h col. $h \ne j$, $h \le n+1$) other than jth column within nth column of (16) to form an $(n+1)\times(n+1)$ matrix (Note that the last column of the $(n+1)\times(n+1)$ matrix is formed by multiplying each element of the finally chosen column by $(K_h^{\ell} \sin \alpha_h^{\ell})/(K_h^{\ell} \cos \alpha_h^{\ell})/(K_h^{\ell} \cos \alpha_h^{\ell})$ alternately.)

where

$$j \le n$$
, $h \ne j$ and $h \le n+1$

Theorem (5-3) is satisfied iff both matrices above are nonsingular. The first matrix is non-singular iff there exists at least one non-zero valued state variable, x_i in (15) such that $\alpha_i^k \neq K\pi/2$ for all k.

Due to its structure, only two cases need be considered to be assured that the second matrix is nonsingular. Namely,

- 1. two adjacent rows containing ω_i will be linearly independent iff neither of the following occurs:
 - a. $\cos \alpha_h / \cos \alpha_j \neq \sin \alpha_h / \sin \alpha_j$
 - b. neither $K_h = 0$ nor $K_j = 0$. This is equivalent to conditions (2) and (3) of the theorem.
- 2. Two nonadjacent rows will be linearly independent iff condition (1) of the theorem is satisfied.

Therefore, it is proven that all the rows of F(t) are linearly independent.

Now to prove linear independence of the columns of F(t), theorem (5-3) in [5] is applied to the transpose of F(t). Thus, if

$$\rho \left[F^{T}(0) : F^{T}(1)(0) : F^{T}(2) : \dots : F^{T}(n)(0) : \dots \right] = n+1$$

where $\rho[\cdot]$ is the rank of $[\cdot]$, and

$$F^{T^{(i)}}(0) = \frac{d^i}{dt^i} F^T(t) \Big|_{t=0}$$

then the columns of F(t) are linearly independent. If it can be shown that $\rho[F^T(0)] = n+1$, then the proof will be completed.

Noting

$$F^{T}(0) = \begin{bmatrix} \kappa_{1}^{1} \sin \alpha_{1}^{1} & \kappa_{2}^{1} \sin \alpha_{2}^{1} & \cdots & \kappa_{n}^{1} \sin \alpha_{n}^{1} & 0 \\ \omega_{1}^{1} \kappa_{1}^{1} \cos \alpha_{1}^{1} & \omega_{1}^{1} \kappa_{2}^{1} \cos \alpha_{2}^{1} & \cdots & \omega_{1}^{1} \kappa_{n}^{1} \cos \alpha_{n}^{1} & \omega_{1}^{1} \\ \vdots & \vdots & & \vdots & & \vdots \\ \frac{m+1}{2} \kappa_{2}^{\frac{m+1}{2}} \cos \alpha_{1}^{\frac{m+1}{2}} & \frac{m+1}{2} \kappa_{2}^{\frac{m+1}{2}} \cos \alpha_{2}^{\frac{m+1}{2}} & \cdots & \frac{m+1}{2} \kappa_{n}^{\frac{m+1}{2}} \cos \alpha_{n}^{\frac{m+1}{2}} & \omega_{n+1}^{\frac{m+1}{2}} \end{bmatrix}$$
and defining

$$\mathbf{F}^{\mathrm{T}}(0) \stackrel{\Delta}{=} [\mathbf{f}_{1} : \underline{\mathbf{f}}_{2} : \cdots \underline{\mathbf{f}}_{i} \cdots \underline{\mathbf{f}}_{n} : \cdots \underline{\mathbf{u}}_{n}]^{\mathrm{T}},$$

it is seen from (15) that

$$\underline{f}_{i} = \left[x_{i}^{1} \dot{x}_{i}^{1} x_{i}^{2} \dot{x}_{i}^{2} \dots \dot{x}_{i}^{\frac{n+1}{2}} \right]^{T} \Big|_{t=0}$$

$$\stackrel{\triangle}{=} \left[f_{1i} f_{2i} \dots f_{ni} f_{n+1} \right]^{T}$$

and

$$\underline{\mathbf{u}} = \begin{bmatrix} 0 & \omega_1 & 0 & \omega_2 & \cdots & \omega_{n+1} \end{bmatrix}^T$$

The proof will be done by contradiction. Thus assuming that $\rho[F^T(0)] < n+1$, it can be stated that

$$\alpha_{1} = \frac{f_{1}}{f_{2}} + \alpha_{2} = 0$$

$$\alpha_{1} = \frac{f_{2}}{f_{2}} + \cdots + \alpha_{n} = 0$$

$$\alpha_{1} = \frac{f_{2}}{f_{2}} + \cdots + \alpha_{n} = 0$$
(18)

for $\underline{\alpha} \not\equiv 0$, where $\underline{\alpha} = [\alpha_1 \alpha_2 \cdots \alpha_{n+1}]^T$.

If (18) is rewritten as

$$\alpha_{1} = \frac{f}{1} + \alpha_{2} = \frac{f}{2} + \dots + \alpha_{n} = -\alpha_{n+1} = \alpha_{n+1} = -\alpha_{n+1} = \alpha_{n+1} = \alpha_{n$$

then there exists an ith row of (19), where i is an odd integer less than n+l such that

$$a_1 f_{i1} + a_2 f_{i2} + \cdots + a_n f_{in} = 0.$$
 (20)

However, since the (i+1)th row was obtained by taking the derivative of the ith row, it follows that

$$\alpha_{1}^{f}_{i+1,1} + \alpha_{2}^{f}_{i+1,2} + \dots + \alpha_{n}^{f}_{i+1,n} =$$

$$\alpha_{1}^{f}_{i1} + \alpha_{2}^{f}_{i2} + \dots + \alpha_{n}^{f}_{in}$$
(21)

To satisfy (21), $\alpha_{n+1} = 0$. Hence (18) becomes

$$\alpha_{1} \stackrel{f}{=} 1 + \alpha_{2} \stackrel{f}{=} 2 + \dots + \alpha_{n} \stackrel{f}{=} n = 0$$
 for $\underline{\alpha} \neq 0$. (22)

Since all the states are assumed to be linearly independent, (22) shows a contradiction. Therefore, it is proven that all the columns of F(0) are linearly independent and $\rho[F^T(0)] = n+1$. Thus, by the comment in Sec.III it follows that the time varying matrix, F(t), is nonsingular for all

tε[0,∞].//

Considering a multi-input system having m inputs, (10) becomes

$$\Phi x + \Psi u = 0 \tag{23}$$

where Φ , Ψ , x, u are defined in II, and (11) becomes

$$\sum_{i=1}^{n} \phi_{ij} x_{j} + \sum_{k=1}^{m} \psi_{ik} u_{k} = 0$$
 (24)

where ϕ and ψ are the elements of Φ and Ψ matrices respectively. Applying the superposition property to (23),

$$\sum_{j=1}^{n} \phi_{ij} x_{j}^{x} + \psi_{ik} u_{k} = 0 , \qquad k=1,2,...n$$
(25)

(24) is basically similar to (11), and an extension of the previous theorem to multi-input system follows.

Corollary:

In multi-input adaptive systems, the convergence of Φ , Ψ matrices to the null is guaranteed if

1. each input contains at least q distinct frequencies, where
q = N.U.I.[R/2], in which

$$R = U.B.[R_i^k]$$

 $R_i^k \stackrel{\Delta}{=}$ number of nonzero terms in (25), i=1,2,...n, k=1,2,...m.

2. conditions 2) and 3) of the main theorem are satisfied with respect to each input. 7

Proof:

When extended to the multi-input case, (11) takes the form of (24). However, due to the multi-input feature, it follows that each input should have a different set of q distinct frequencies so that superposition property can be applied to each term of $\sum_{k=1}^{m} \psi_i k u_k$ in (24) separately, thereby guaranteeing that each $\psi_i \rightarrow 0$.

Assuming that each u_k , k=1,2,...m, contains the required number of distinct frequencies, u_k may be written as follows:

$$u_{k} = \sum_{\ell=1}^{q} u_{k}^{\ell} \tag{26}$$

Here again, using the superposition property, we have

$$\sum_{j=1}^{n} \phi_{\hat{i}j} x_{j}^{\ell} + \psi_{\hat{i}k} u_{k} = 0$$
 (27)

where $l=1,2,\ldots,q$,

and the \hat{i}^{th} row is the row of the maximum number of nonzero terms. (27) is identical to (13) and it has R terms where R is defined in the corollary. Using the same procedure as in the case of the single input system, the convergence of $\psi_{\hat{i}\hat{k}}$ and $\phi_{\hat{i}\hat{j}}$, $j \in R$, can be proved. Consequently $[\gamma_i] \equiv 0$ for all i.//

⁷The corollary can be extended as in [2] to require only that [A,B] be controllable.

In order to interpret the above results and notations used in the preceding development, an illustrative example will be worked out. However, it is noted that application of the theorem can be made directly upon determination of R.

Example:

Consider a fourth order system with two inputs. Assume from (23) that the Φ , Ψ matrices take the following forms.

$$\Phi = \begin{bmatrix} \phi_{11} & 0 & 0 & 0 & 0 \\ \phi_{21} & \phi_{22} & 0 & 0 & 0 \\ 0 & 0 & 0 & \phi_{34} \\ \phi_{41} & \phi_{42} & 0 & 0 \end{bmatrix} \quad \text{and} \quad \Psi = \begin{bmatrix} \Psi_{11} & 0 & 0 & 0 \\ \Psi_{21} & \Psi_{22} & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix}$$

From (25) and the condition (1) of the corollary, we have

$$R = U.B.[R_1^1, R_1^2, R_2^1, R_2^2, R_3^1, R_3^2, R_4^1, R_4^2]$$

$$= U.B.[2, 1, 3, 3, 1, 2, 1, 1]$$

Therefore,

$$q = N.U.I.[R/2]$$
 $= N.U.I.[3/2]$
 $= 2$

and by the theorem two frequencies are called for.

Since the second row is identified with the largest value of R, it is used as a starting point. Thus, with two distinct frequencies, ω_1 and ω_2 , (27) can be written for the second row of (23) with either u_1 or u_2 equal to zero. Choosing $u_1 = \sin \omega_1 t + \sin \omega_2 t$, $u_2 \equiv 0$,

$$\omega_{1}: \phi_{21}x_{1}^{1} + \phi_{22}x_{2}^{1} + 0 \cdot x_{3}^{1} + 0 \cdot x_{4}^{1} + \psi_{21}u_{1}^{1} = 0$$

$$\omega_{2}: \phi_{21}x_{1}^{2} + \phi_{22}x_{2}^{2} + 0 \cdot x_{3}^{2} + 0 \cdot x_{4}^{2} + \psi_{21}u_{1}^{2} = 0$$
(28)

The time derivative of (28), noting that $\dot{\phi}_{21} = \dot{\phi}_{22} = \dot{\psi}_{21} = 0$ in the steady state, is

$$\phi_{21}\dot{x}_{1}^{1} + \phi_{22}\dot{x}_{2}^{1} + \psi_{21}\dot{u}_{1}^{1} = 0$$

$$\phi_{21}\dot{x}_{1}^{2} + \phi_{22}\dot{x}_{2}^{2} + \psi_{21}\dot{u}_{1}^{2} = 0$$
(29)

(28) and (29) take the form of (15). Since only three independent equations are needed, it suffices to use (28) and the first equation of (29). Accordingly,

$$\begin{bmatrix} \vec{x}_{1}^{1} & \vec{x}_{2}^{1} & \vec{u}_{1}^{1} \\ \dot{x}_{1}^{1} & \dot{x}_{2}^{1} & \dot{u}_{1}^{1} \\ \dot{x}_{2}^{2} & \vec{x}_{2}^{2} & \vec{u}_{1}^{2} \end{bmatrix} \begin{bmatrix} \phi_{21} \\ \phi_{22} \\ \psi_{21} \end{bmatrix} = 0$$
(30)

The first matrix of (30) can be written as

$$F(t) = \begin{bmatrix} \kappa_{1}^{1} \sin (\omega_{1}t + \alpha_{1}^{1}) & \kappa_{2}^{1} \sin (\omega_{1}t + \alpha_{2}^{1}) & \sin \omega_{1}t \\ \omega_{1}\kappa_{1}^{1} \cos (\omega_{1}t + \alpha_{1}^{1}) & \omega_{1}\kappa_{2}^{1} \cos (\omega_{1}t + \alpha_{2}^{1}) & \omega_{1} \cos \omega_{1}t \\ \kappa_{1}^{2} \sin (\omega_{2}t + \alpha_{1}^{2}) & \kappa_{2}^{2} \sin (\omega_{2}t + \alpha_{2}^{2}) & \sin \omega_{2}t \end{bmatrix}$$

corresponding to F(t) in (15).

Now form (16) up to the second derivative of F(t) and set $t = t_0 \equiv 0$. By using the rule described in II, and choosing the appropriate columns we have

$$\begin{bmatrix} \vec{K}_{1}^{1} \sin \alpha_{1}^{1} & -\omega_{1}^{2} \vec{K}_{1}^{1} \sin \alpha_{1}^{1} & \vec{K}_{2}^{1} \sin \alpha_{2}^{1} \\ \omega_{1} \vec{K}_{1}^{1} \cos \alpha_{1}^{1} & -\omega_{1}^{3} \vec{K}_{1}^{1} \cos \alpha_{1}^{1} & \omega_{1} \vec{K}_{2}^{1} \cos \alpha_{2}^{1} \\ \vec{K}_{1}^{2} \sin \alpha_{1}^{2} & -\omega_{2}^{2} \vec{K}_{1}^{2} \sin \alpha_{1}^{2} & \vec{K}_{2}^{2} \sin \alpha_{2}^{2} \end{bmatrix}$$
(31)

Factoring (31) into two matrices as shown in (17), the following expression is obtained:

$$\begin{bmatrix} \kappa_{1}^{1} \sin \frac{1}{1} & & & \\ & \kappa_{1}^{1} \cos \alpha_{1}^{1} & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\$$

It has been shown in II that these two matrices are nonsingular under the conditions described in the corollary. Therefore, from (30)

$$\begin{bmatrix} \phi_{21} \\ \phi_{22} \\ \psi_{21} \end{bmatrix} \equiv 0 \tag{32}$$

Considering the second row of (23) with input u_2 acting (i.e. $u_1 \equiv 0$), and using (32), it follows

$$\psi_{22}u_{2} = 0 \tag{33}$$

and

$$\psi_{22} = 0 \tag{34}$$

Now consider the first row of (23) with $u_1 \not\equiv 0$, $u_2 \equiv 0$. Since this row has only two nonzero elements of Φ , only one of the two distinct frequencies used previously need be employed here. Choosing ω_1 , we have

$$\omega_{1} \colon \phi_{11} x_{1}^{1} + \psi_{11} u_{1}^{1} = 0 \tag{35}$$

Noting that $\dot{\phi}_{11} = \dot{\psi}_{11} = 0$ in the steady state, we have for the time derivative of (35)

$$\phi_{1} \dot{x}_{1}^{1} + \psi_{11} \dot{u}_{1}^{1} = 0 \tag{36}$$

The matrix form of (35) and (35) become

$$\begin{bmatrix} \mathbf{x}_{1}^{1} & \mathbf{u}_{1}^{1} \\ \dot{\mathbf{x}}_{1}^{1} & \dot{\mathbf{u}}_{1}^{1} \end{bmatrix} \begin{bmatrix} \phi_{11} \\ \psi_{11} \end{bmatrix} = 0$$
 (37)

Rewriting the first matrix of (37),

$$F(t) = \begin{bmatrix} K_1^1 \sin (\omega_1 t + \alpha_1^1) & \sin \omega_1 t \\ \omega_1 K_1^1 \cos (\omega_1 t + \alpha_1^1) & \omega_1 \cos \omega_1 t \end{bmatrix}$$

Now form (16), i.e. set $t = t_0 = 0$, and make use of the rule described in II to obtain the following expression:

$$\begin{bmatrix} K_1^1 \sin \alpha_1^1 & 0 \\ \omega_1 K_1^1 \cos \alpha_1^1 & \omega_1 \end{bmatrix}$$
(38)

The factored form of (38) becomes

$$\begin{bmatrix} K_1^1 \sin \alpha_1^1 & 0 \\ 0 & K_1^1 \cos \alpha_1^1 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ \omega_1 & \omega_1/K_1^1 \cos \alpha_1^1 \end{bmatrix}$$
(39)

showing that these two matrices are nonsingular under the required conditions.

Therefore, from (37)

$$\begin{vmatrix} -\phi_{11} \\ \psi_{11} \end{vmatrix} = 0 \tag{40}$$

A similar exercise can be applied to the other rows with the result that all the rest of the parameter differences converge to zero.

IV. Computer Simulation

Computer simulation was undertaken to test the sufficiency aspect of the theorem. The various cases listed in Table 1 were studied, as discussed below.

- 1) In [4], an example of a 4th order multivariable system with two inputs and three unknown parameters was simulated in an identification scheme. The input signals consisted of periodic square waves of unit height with fundamental frequencies of ω_1 = 1 rad/sec. and ω_2 = 2 rad/sec for inputs u_1 and u_2 respectively. For this same example, it is found that R = 1. Thus, by the corollary it suffices that each input be excited by a single frequency and that these frequencies be distinct. The inputs chosen were u_1 = $\sin \omega_1 t$ and u_2 = $\sin \omega_2 t$, where ω_1 = 1 rad/sec. and ω_2 = 2 rad/sec. The results shown in Fig. 1-a, b, c for parameters, a_{31}^* , a_{33}^* and b_{21}^* respectively demonstrate that, according to the theorem, one frequency for each input is sufficient.
- 2) Using the identification technique developed in [4], a plant of second order with a single input and two unknown parameters were considered. The input frequency was selected such that all the states in the plant should encounter $K\pi/2$ phase shift. The selected frequency was 4 rad./sec. As shown in Fig. 2a and b, the values of the unknown plant parameters were identified showing that the phase shift requirement is not a necessary condition.

The plant and model were as follows:

$$A^{*} = \begin{bmatrix} 0 & \overline{1} \\ -16 & -8 \end{bmatrix} \qquad b^{*} = \begin{bmatrix} 0 & \overline{1} \\ 16.0 \end{bmatrix}$$

$$A = \begin{bmatrix} 0 & 1 \\ a_{21} & a_{22} \end{bmatrix} \qquad b = \begin{bmatrix} \overline{0} \\ 16 \end{bmatrix} \qquad C = \begin{bmatrix} 0 & 1 \\ -9 & -6 \end{bmatrix}$$

where $a_{21}(0) = -25.0$ $a_{22}(0) = -10.00$

$$P = \begin{bmatrix} 75.75 & 10 \\ 10 & 1.75 \end{bmatrix} \qquad Q = \begin{bmatrix} 180 & 0 \\ 0 & 1 \end{bmatrix}$$

- 3) Using the same system as in 2), the number of unknown parameters was increased from two to three, such that R = 3. Therefore, two distinct frequencies are called for. To test the theorem, only one frequency was provided. Fig. 3a-f show that the model parameters did not converge to the plant parameters, and, furthermore, steady-state values of model parameters depended upon the size of input. Accordingly it is conjectured that the frequency requirement is a necessary condition.
- 4) It has been conjectured that the controllability requirement is not a necessary condition. To verify the statement above, the following case of an uncontrollable third-order plant with three unknown parameters was simulated.

$$A^* = \begin{bmatrix} 0 & 0 & -1\overline{2} \\ 1 & 0 & -19 \\ 0 & 1 & -8 \end{bmatrix} \qquad B^* = \begin{bmatrix} 2 \\ 3 \\ 1 \end{bmatrix}$$

$$C = \begin{bmatrix} -10 & 0 & 0 \\ 0 & -12 & 0 \\ 0 & 1 & -15 \end{bmatrix} \qquad A = \begin{bmatrix} 0 & 0 & -12 \\ a_{21} & 0 & a_{23} \\ 0 & 1 & a_{33} \end{bmatrix}$$

B = B*

where $a_{21}(0) = 3.0$ $a_{23}(0) = -15.0$ and $a_{33}(0) = -6.0$ $\omega_{1} = 5.5 \text{ rad/sec}$

and

$$P = \begin{vmatrix} \overline{5} & 0 & \overline{0} \\ 0 & 10 & 0 \\ 0 & 0 & 10 \end{vmatrix} \qquad Q = \begin{vmatrix} \overline{100} & 0 & 0 \\ 0 & 240 & 0 \\ 0 & 0 & 300 \end{vmatrix}$$

As shown in Fig. 4a, b, c, the convergence of the model parameters to the plant parameters is accomplished in spite of the fact that the system is uncontrollable. This example proves the controllability is not a necessary condition for parameter convergence.

5) Due to the fact that the theorem has been developed in state variable rather than transfer function form, in some cases, less frequencies are required than the number of frequencies called for by [1], [2]. The example below illustrates this point, wherein

PLANT:
$$A^* = \begin{bmatrix} 0 & -6 \\ 1 & -5 \end{bmatrix}$$

$$B^* = \begin{bmatrix} 1 \\ 2 \end{bmatrix}$$
MODEL:
$$C = \begin{bmatrix} -10 & 0 \\ 0 & -12 \end{bmatrix}$$

$$A = \begin{bmatrix} 0 & -6 \\ 1 & a_{22} \end{bmatrix}$$

$$B = \begin{bmatrix} b_1 \\ b_2 \end{bmatrix}$$

where $a_{22}(0) = -8.0$, $b_1(0) = 3.0$, $b_2(0) = 5.0$ and

$$P = \begin{vmatrix} \overline{5} & \overline{0} \\ \underline{0} & \underline{10} \end{vmatrix} \qquad \qquad Q = \begin{vmatrix} \overline{-100} & 0 \\ \underline{0} & -240 \end{vmatrix}$$

Since R = 2, one frequently is called for. Setting ω_1 = 4.5 rad/sec, simulation shows that the model parameters converged to the plant parameters [See Fig. 5 a,b,c].

In order to apply the theorems in [1] and [2], the plant state equation is first put in form

$$x + \bar{a}_1 \dot{x} + \bar{a}_0 x = \bar{b}_0 u + \bar{b}_1 \dot{u}$$

where

$$\overline{a}_1 = -a_{22}$$
 $\overline{a}_0 = 6.0$
 $\overline{b}_0 = -a_{22}b_1 - 6b_2$
 $\overline{b}_1 = b_1$

Since the number of unknown parameters is three, according to [1] and [2], two frequencies are called for.

V. Conclusions

The theorem derived in this paper provides sufficient conditions regarding the frequency content of the inputs to a Liapunov-adaptive identification process, so as to guarantee nullification of the parameter-error vector. As such it complements the several other similar theorems which have appeared in the literature [1],[2],[3]. In contrast to previous work, however, this theorem utilizes knowledge of specific locations of unknown parameters in the A, B matrices, thereby in some instances significantly reducing the frequency requirements.

Because this theorem, as well as others, provide sufficient conditions only, several simulations have been made to test whether these conditions may in fact not be necessary. It is found, notably, that the controllability condition is not necessary, while the number of frequencies required appears to be a necessary condition. As summarized in Table 1, other sufficiency conditions are tested as well.

These results point to the need for deriving a more complete theorem than has yet been reported, in which both necessary and sufficient conditions are established.

Figure	n	m	q	Condition	Result
Fig. 1 a,b,c	4	2	1	Ref. [4] Example	Converged
Fig. 2 a,b	2	1	1	Violation of Kπ/2 Phase Shifts	Converged
Fig. 3 a-f	2	1	2	Number of Frequencies less than q	Not Converged
Fig. 4 a,b,c	3	1	1	Uncontrollable plant	Converged
Fig. 5 a,b,c	2	1	1	Number of frequencies less than in Ref. [1],[2]	Converged

TABLE 1

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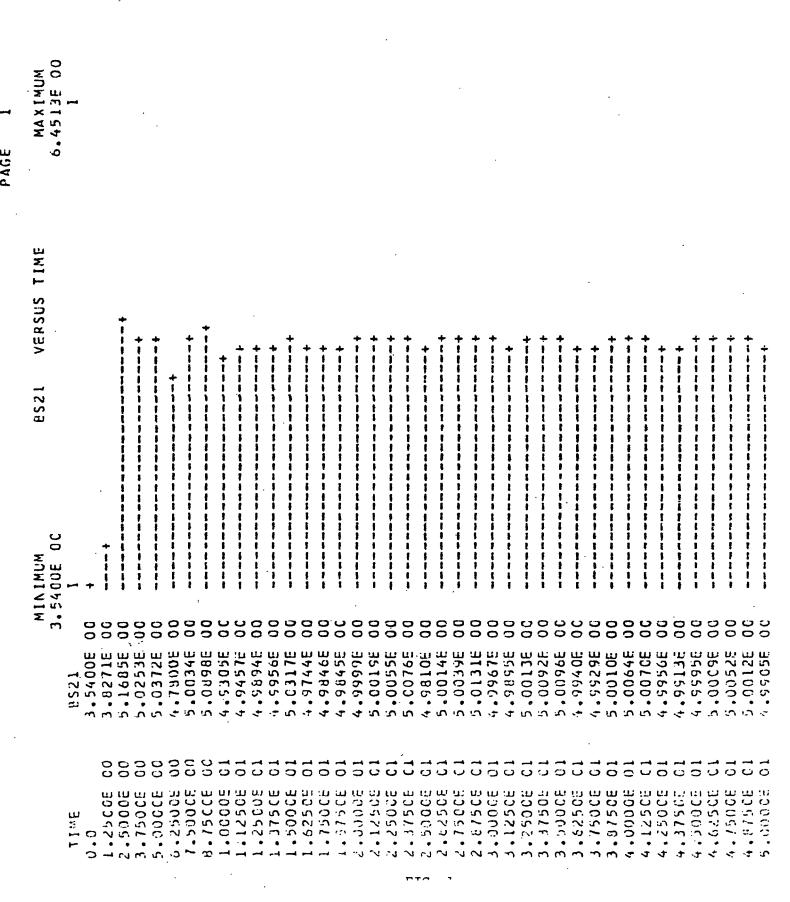
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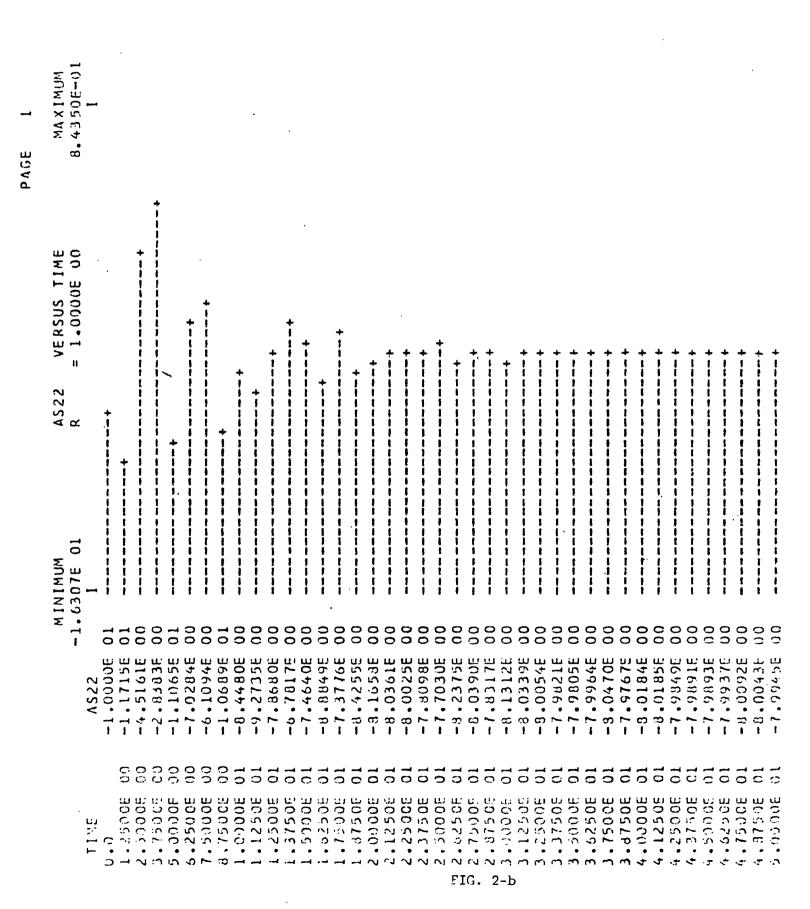
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Note: Input Amplitude = 1.0

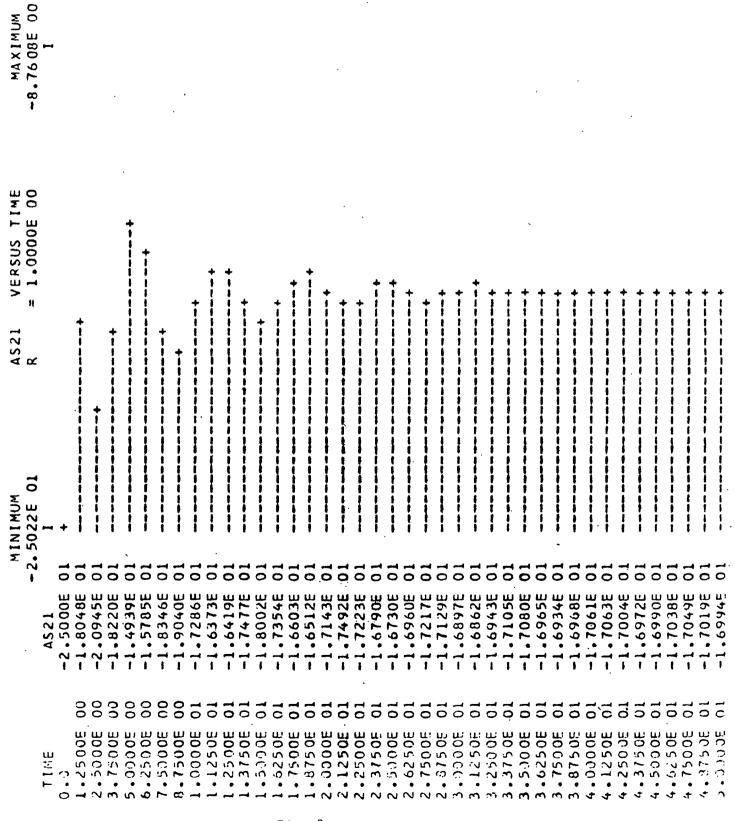
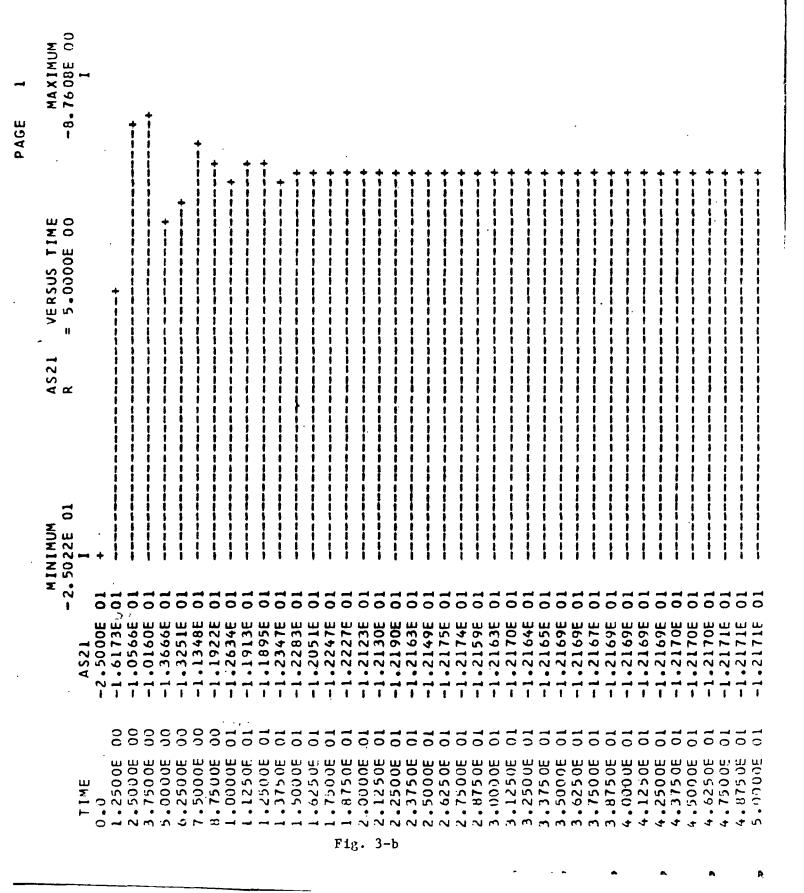
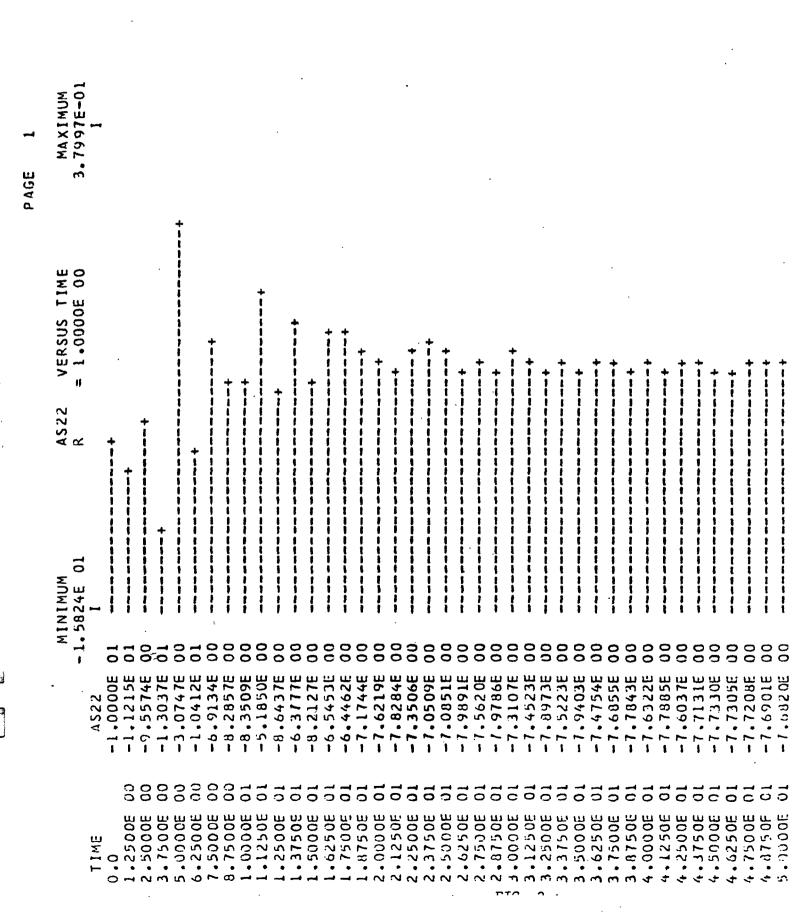


Fig. 3-a

Note: Input Amplitude = 5.0



Note: Input Amplitude = 1.0

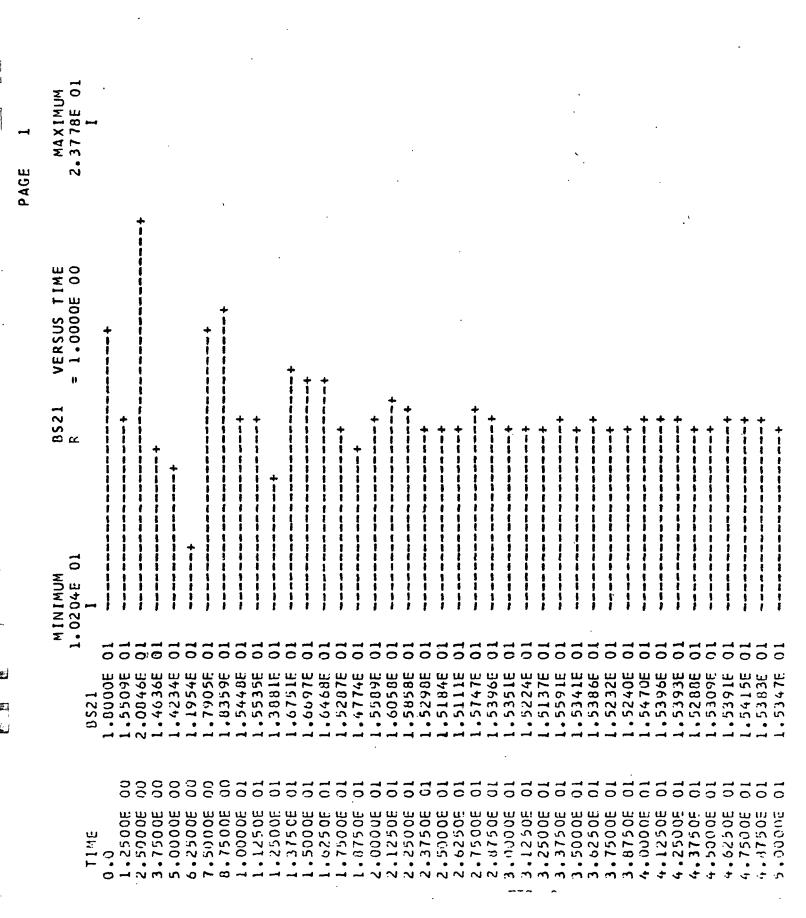


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Note: .Input Amplitude = 1.0



Note: Input Amplitude = 5.0

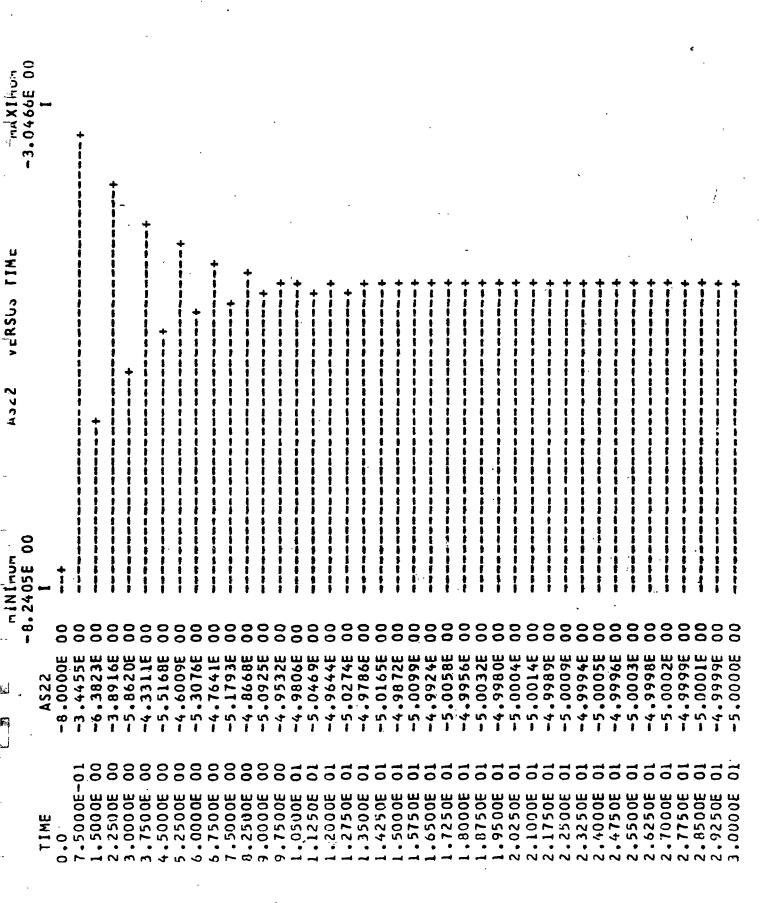
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